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Implementation of a High Efficiency Grid-Tied Multi-Level Photovoltaic Power Conditioning System Using Phase Shifted H-Bridge Modules

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Abstract

This paper proposes a high efficiency three-phase cascaded phase shifted H-bridge multi-level inverter without DC/DC converters for grid-tied multi string photovoltaic (PV) applications. The cascaded H-bridge topology is suitable for PV applications since each PV module can act as a separate DC source for each cascaded H-bridge module. The proposed phase shifted H-bridge multi-level topology offers advantages such as operation at a lower switching frequency and a lower current ripple when compared to conventional two level topologies. It is also shown that low ripple sinusoidal current waveforms are generated with a unity power factor. The control algorithm permits the independent control of each DC link voltage with a maximum power point for each string of PV modules. The use of the controller area network (CAN) communication protocol for H-bridge multi-level inverters, along with localized PWM generation and PV voltage regulation are implemented. It is also shown that the expansion and modularization capabilities of the H-bridge modules are improved since the individual inverter modules operate more independently. The proposed topology is implemented for a three phase 240kW multi-level PV power conditioning system (PCS) which has 40kW H-bridge modules. The experimental results show that the proposed topology has good performance.

Key words: H-bridge multilevel inverter, Photovoltaic PCS (Power Conditioning System)

I. INTRODUCTION

In recent years, renewable energy resources have become an increasingly important part of power generation. Aside from assisting in the reduction of the emission of greenhouse gases, they add much needed flexibility to the energy resource mix by decreasing the dependence on fossil fuels. Due to their modular characteristics and ease of installation and because they can be located closer to users, PV systems have great potential as a distributed power source [1]. The focus is to make use of abundantly available PV energy through the design and control of an inverter suitable for photovoltaic applications.

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Many topologies for large scale PV generation systems have been proposed in the literature [2]-[7]. Fig. 1 shows the general topology of a large scale PV generation system. The central PV generation system is shown in Fig.1 (a). It is widely used in large scale PV generation systems because of its cost effectiveness. However, it has a drawback under shading conditions. Fig 1(b) shows the string topology which consists of series-connected PV strings with a general DC/AC inverter. The string PCS can avoid most of the drawbacks of the central PCS. The generation efficiency is higher than that of a central generation system due to its ability to adopt a MPPT controller to each individual string. The mismatch problem of parallel-connected strings is avoided under shading conditions. However, it has a power limit due to its limited number of series connections in order to increase the nominal power. To avoid this problem, the multi-string topology was proposed as shown in Fig. 1(c). It consists of several PV strings that connect DC/DC converters to a general DC/AC inverter. The multi-string PCS has the advantages of the central PCS and the string PCS: no

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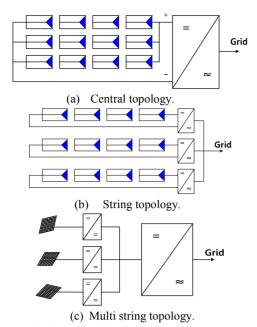


Fig. 1. Topologies of large scale PV generation system.

mismatch problem of the parallel-connected strings, no partial shading problem, no nominal power limitation, etc [21]. However, it is a two stage power conversion topology which means low efficiency. To increase the power conversion efficiency, a novel multi string PCS topology is needed.

These days, multi-level inverter topologies have been considered for photovoltaic applications. A multi-level inverter can generate low harmonic voltage waveforms with a low switching frequency. Additionally, the multi-level topology has several DC links which makes it possible to control the voltage independently with MPPT in each string. In addition, this characteristic makes it possible to increase the PV generation efficiency in the case of partial shading due to unbalanced solar radiation. Among multi-level topologies, the cascaded phase shifted H-bridge topology has several advantages such as a PV PCS. That is, an output voltage synthesis with a higher number of levels. This number of levels grows according to the number of series-connected modules in the converter. This topology can make a modular design and has power semiconductors with a lower rating than the conventional two level topology. This leads to cost savings. In [8][9], a single phase cascaded H-bridge multi-level inverter has been presented. However, it does not describe a large scale three phase PV PCS system.

This paper proposes a high efficiency cascaded H-bridge multi-level topology for large scale grid-tied three-phase multi-string PV systems. The use of MPPT algorithms that independently control the DC link voltages is applied for the control of the three-phase cascaded H-bridge multi-level converter. In addition, The use of the controller area network (CAN) communication protocol for H-bridge multi-level inverters, along with localized PWM generation and PV

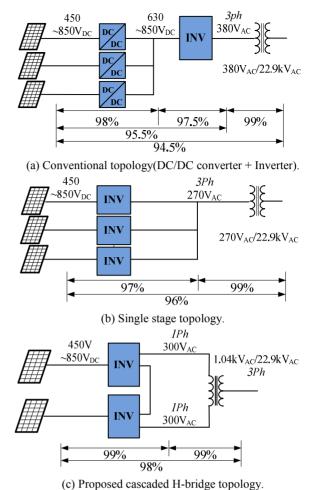


Fig. 2. PV PCS schemes for multi string topology.

voltage regulation are implemented. It is also shown that the expansion and modularization capabilities of the H-bridge modules are improved since the individual inverter modules operate more independently. The validity of the proposed system is verified experimentally on a three-phase 240kW cascaded H-bridge multi-level PV PCS.

II. CHARACTERISTICS OF MULTI-LEVEL INVERTERS FOR PV APPLICATIONS

The multi-level inverters for PV applications have several dc links, making possible independent voltage control and tracking of the maximum power point (MPP) in each string.

The characteristics of the multi-level inverters for PV applications can increase the generation efficiency of the PV system in the case of a mismatch in the string, due to unequal solar radiation, aging of the PV modules, different types of cells, or the accumulation of dust on the surface of the modules [10]-[16]. The cascaded H-bridge multi-level inverter is an interesting conversion topology among the available multi-level inverters because of its modularity, simple layout, and reduced number of components. The major drawback of this topology is the large number of distinct dc links. However, it is suitable for multi-string PV

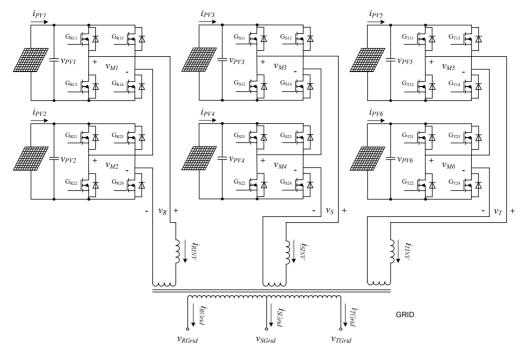


Fig. 3. Proposed three phases cascaded H-bridge PV PCS.

systems as a PV PCS.

The conventional PV PCS scheme for the multi-string topology is shown in Fig. 2(a). It consists of a DC/DC converter for each of the PV modules and the grid-tied inverter. If it is supposed that the efficiency of the DC/DC converters is 98%, the efficiency of the inverter is 97.5%, and the efficiency of the transformer is 99%, then the efficiency of the system is about 94.5% [23].

The conventional scheme has a low efficiency and is not a suitable topology for large PV generation systems due to its two stage conversion topology. Fig. 2(b) shows a single stage multi-string topology with a parallel-connected inverter. The conversion efficiency is higher than the conventional system due to its single conversion. However, it has a power limit due to its limited number of series connections in order to increase the nominal power. To overcome this problem, a novel multi string PCS is in need of a high efficiency and effective cascaded H-bridge inverter topology without a DC/DC converter, as shown in Fig. 2(c). This topology has only cascaded inverters with a high output voltage in each The proposed transformer 1.04KV/22.9KV with a high efficiency.

As a result, the efficiency of this topology achieves 98%. Nevertheless, it is in need of an independent MPPT control algorithm with grid current control. The proposed topology is presented in the next section. The advantages of the proposed multi-string construction using the H-bridge topology are listed as follows: 1) 5-level inverter output voltage; 2) reduced switching loss; 3) small size filter inductor and transformer; 4) low THD current waveforms; and 5) better efficiency than the conventional multi-level topology.

III. PROPOSED TOPOLOGY DESCRIPTION

Fig. 3 shows that the proposed three phase cascaded H-bridge multi-string PV PCS uses a multi-level topology which has an independent MPPT operation for each of the PV modules [17]. The proposed topology has a 5-level PWM switching waveform which reduces the switching frequency better than the conventional topology.

A. PWM Scheme with the Phase Shift Unipolar Method

Many PWM techniques have been proposed over years for multi-level inverters [18]. Among these, the methods based on sine-triangular comparison take a great part, including carrier phase-shifted PWM schemes (PS-PWM), carrier disposition PWM schemes (CD-PWM), and compound carrier disposition PWM schemes (CCD-PWM). The PS unipolar sinusoidal PWM (PS SPWM) switching schemes, which are used in this paper, are described with the aid of Fig. 4 for a two layer H-bridge multi-level inverter.

The multi-carrier PWM method is based on a comparison of a sinusoidal reference waveform with vertically shifted carrier signals. The method uses N-1 level carrier signals to generate the N-level inverter output voltage. In multi-level inverters, the frequency modulation index, m_f and the amplitude modulation index, m_a , are defined as:

$$m_f = \frac{f_{car}}{f_{ref}} \tag{1}$$

$$m_f = \frac{f_{car}}{f_{ref}}$$

$$m_a = \frac{A_{ref}}{(N-1)A_{car}}$$
(2)

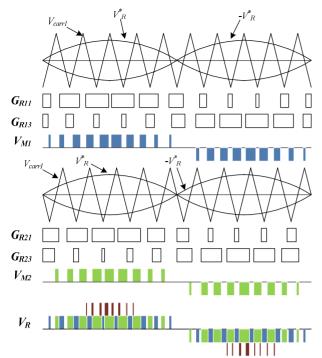


Fig. 4. Phase shifted unipolar PWM scheme for the proposed multi-level inverter.

Where A_{car} is the amplitude of the carrier signal; f_{car} is the frequency of the carrier signal; A_{ref} is the amplitude of the reference signal; and f_{ref} is the frequency of the reference signal. The H-bridge topology has a left leg and a right leg. The frequency of the output voltage waveform of the H-bridge topology is double each leg of the PWM switching frequency. The phase shift between the upper H-bridge and the lower H-bridge is 90 degree. It could be double the switching frequency of each H-bridge. In addtion, the output

voltage of the two H-bridge topology is a 5-level waveform (V_R) as shown in Fig. 4 $(m_f=35, m_a=0.8)$. One of the main advantages of this PWM scheme is that the harmonics of the resultant H-bridge multi-level inverter output voltage only appear as sidebands centered around the frequency of 2Nfs and its multiples (N is the number of single-phase inverters, and fs is the frequency of the triangle carrier signals). Therefore the H-bridge multi-level inverter output voltage has a very high equivalent switching frequency, even if the switching frequency of the individual switches is not so high.

B. Control Scheme

The structure of the multi-level inverter for the PV PCS has various PV modules but a single phase current output. Therefore, it can not apply a general MPPT algorithm [13]-[15]. The control algorithm for the proposed topology controls the output current and several of the input voltages simultaneously [8][9]. One of them is used to adjust the capacitor voltage in each of the dc links, and the other one is necessary for the generation of a sinusoidal input current with a unity power factor.

Fig. 5 shows the control algorithm of the proposed n-level H-bridge inverter with a single phase. The P&O (Perturb & Observe) method applies the MPPT algorithm. The output of MPPT function block is the input reference voltage ($V_{PV1}^*, V_{PV2}^*, ..., V_{PVn}^*$), where V_{PVn}^* is the input reference voltage of the n-th H-bridge, and V_{PV}^* is the sum of the input reference voltages of the H-bridges. The modulation index of the n-th H-bridge (V_{CONn}) is the output of the voltage controller directly.

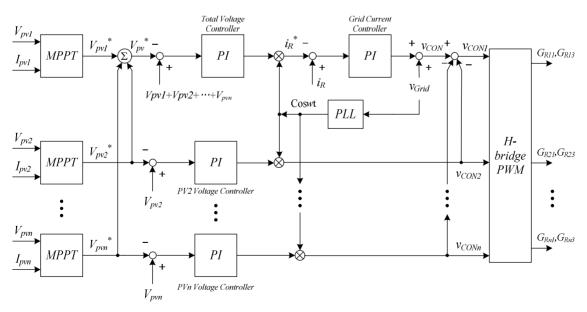


Fig. 5. Control scheme per phase of the proposed topology.

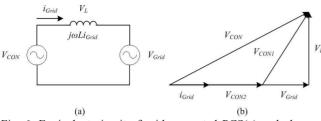


Fig. 6. Equivalent circuit of grid-connected PCS(a) and phasor diagram(b).

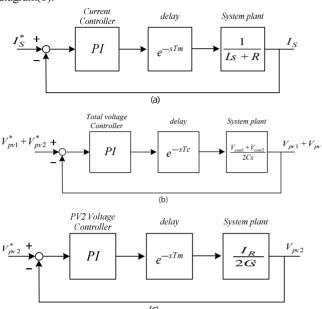


Fig. 7. Control loop (a) Current (b) Total voltage (c) V_{pv2} voltage.

The reference PV voltage (V_{PV}^*) is the sum of all of the outputs of the MPPT function block, and the output of the voltage controller of the first H-bridge is the reference current controller. The modulation index of the first H-bridge (V_{CON1}) subtracts the total of each of the voltage controllers $(V_{CON1}, V_{CON2}, ..., V_{CONn})$ from the output of the current

TABLE I
CONTROLLER PARAMETERS

Controller	P gain	I gain
Grid controller	0.06	0.0035
Total voltage	0.3	0.002
V_{PV2} voltage	0.27	0.003

controller (V_{CON}).

The power factor can be controlled by changing the magnitude and phase of the voltage of each of the PV cells, as shown Fig. 6. A unity power factor is obtained when a voltage V_L with a phase of 90°, with respect to the grid voltage, is generated.

C. Controller Design

In this control scheme, two PI controllers are necessary for independent control of each of the dc-link voltages and one PI controller is necessary for grid current control. The tuning procedure for the three control loops in Fig. 7 is described [8][9].

Fig. 7(a) shows the simplified control scheme of the current control loop. The dymanic of the current loop is much faster than the voltage loop.

The current controller was designed using a bandwidth of 400rad/s which is high enough to provide an adequate current tracking and to filter the harmonics of the modulation. The voltage controllers were designed considering the frequency of the MPPT algorithm, which, in this case, is equal to 5 Hz. For these controllers, a bandwidth between 50 and 200rad/s was considered.

The discrete PI control parameters used in simulation and the experimental results are described in Table I.

D. Implementation of the Proposed Method

In this paper, the PWM signal is generated in each of the individual modules. Therefore, the PWM signals must be synchronized. The PWM of each module is synchronized

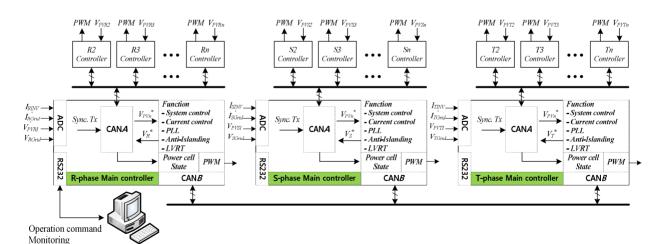


Fig. 8. System Control block diagram of proposed topology.

 $TABLE\ II$ The Specification of 240kW H-bridge multi-level PV PCS

MPPT range (H-bridge module)	400~800Vdc	
Rated Power	240kW	
Inverter output voltage/current	880Vrms/160Arms	
Grid voltage/current	380Vrms/380Arms	
Filter Inductance	600uH	
Switching frequency	2.1KHz	
Transformer	880V(Y)/380V(Δ)	

using the CAN communication receive interrupt. The overall system control configuration and the features of the phase controller, the CAN communication, and the inverter modules are summarized in Fig. 8. The Main phase controller includes the system control, the phase current control, the PLL control, and the anti-islanding control. It also performs state monitoring, diagnosis, and communication functions. The module controller has a maximum power point tracking algorithm using an incremental conductance method. It also controls the phase shift and the required PWM voltage calculated by the main controller. Each module controller is connected to the main controller via a high speed link that is made of an optical cable.

IV. EXPERIMENTAL STUDY

A. System Configuration

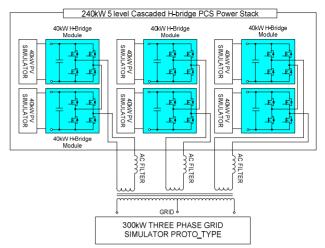
A 240kW three-phase two-layer cascaded H-bridge multi-level inverter has been manufactured for testing purposes, as shown in Fig. 9(a). The two modules per output phase can be seen in a photograph of the hardware used for the experimental evaluation in Fig. 9(b).

B. Experimental Results

Fig. 10 shows the inverter output and the PWM waveforms of the upper/lower H-bridge modules. The proposed topology consists of two H-bridge modules per phase. The inverter output voltage is a 5 level waveform.

Fig 11(a) shows that the output voltage waveform of the H-bridge topology is double each leg of the PWM switching frequency. The phase shift between the upper H-bridge and the lower H-bridge is 90 degrees. It can be twice the switching frequency of each H-bridge. In addition, the output voltage of the two H-bridge topology is a 5-level waveform, as shown Fig. 11(b).

Fig. 12 shows the tracking of the MPP curve under different operating points of the PV module. The maximum power point of PV1 is 14.8KW and that of PV2 is 30.28KW. The independent MPPT control algorithm works properly under different operation conditions.



(a) Schematic of 240kW H-bridge 5-level PV PCS.



Fig. 9. Configuration of 240kW Proposed topology.

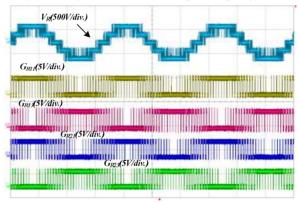


Fig. 10. The inverter output voltage and PWM waveforms (5ms/div.)

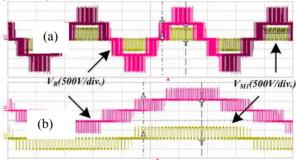


Fig. 11. The H-bridge output voltage waveforms(5ms/div.).

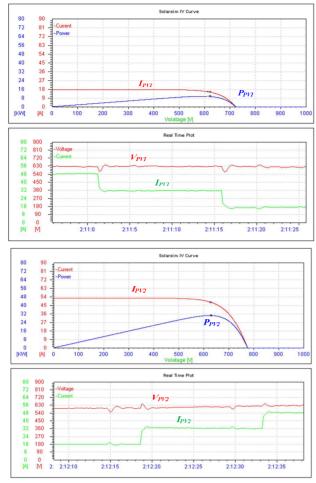


Fig. 12. MPP curve of each H-bridge PV module(PV₁, PV₂).

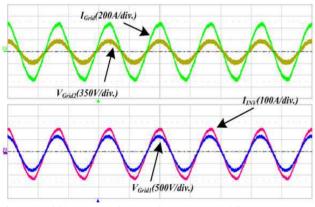


Fig. 13. Grid/Inverter voltage and current(10ms/div.).

Fig. 13 shows the experimental waveforms of the grid voltage and output current of the PCS at 240kW PCS output. The THDi of the output current of the PCS is 2.5% at 3.8% THD ν of the grid voltage. The power factor is 99.8%.

Fig. 14 shows the efficiency curves with respect to the input voltage (400~800Vdc) and the load condition (10~100%). The peak efficiency of the proposed PCS is 99.1% without the grid transformer (98% with the grid transformer).

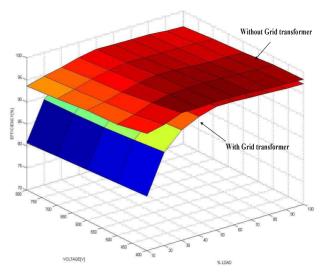


Fig. 14. Efficiency curve of the proposed PCS.

V. CONCLUSIONS

In this paper, a novel topology has been proposed for a PV PCS using a 5-level Phase shifted H-bridge topology with a very high efficiency. The proposed topology greatly reduces the current THD and the filter size while increasing the system efficiency. European system efficiency is up to 97.8% and peak efficiency is 99.1%. The control algorithm for the proposed topology controls the output current with a unity power factor, and the upper and lower PV input voltages by adding a MPPT algorithm simultaneously. In addition, control and state monitoring among each of the phase controllers is implemented using the controller area network (CAN) communication protocol. It has been shown that there is an improvement in the expansion and modularization capabilities in the multi-level topology. The experimental results show the good performance of the proposed topology.

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