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Digital Modeling of a Time delayed Continuous-Time System

시간 지연 연속 시간 시스템의 디지털 모델링

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요 약 연속시간 시스템의 제어 이론은 잘 개발되어 왔다. 컴퓨터 기술의 발달로 인해 디지털 제어 기법이 여러 분야에 적용되어 왔다. 제어 시스템에 시간 지연이 있는 경우는 시스템을 효율적으로 제어하는 것이 어렵다. 제어기와 액추에이터 그리고 센서와 제어기 간에 있는 지연은 제어 성능을 떨어뜨리고 전체 시스템을 불안정하게 할 수 있다. 본 논문에서는 다중의 상태, 입력 그리고 출력 지연을 가지는 제어 시스템을 위한 새로운 근사 이산화 방법과 디지털 설계 그리고 조정 가능한 계수를 가지는 일반화된 쌍선형 변환 방법을 제안한다. 이 방법은 정수의 시간 지연을 가지는 이산 시간 모델을 동일한 연속 시간 모델로 다시 변환할 수 있다. 실제적인 예제를 통해 제안된 방법의 효율성을 증명한다.

Abstract Control Theory for continuous-time system has been well developed. Due to the development of computer technology, digital control scheme are employed in many areas. When delays are in control systems, it is hard to control the system efficiently. Delays by controller-to-actuator and sensor-to-controller deteriorate control performance and could possibly destabilize the overall system. In this paper, a new approximated discretization method and digital design for control systems with multiple state, input and output delays and a generalized bilinear transformation method with a tunable parameter are also provided, which can re-transform the integer time-delayed discrete-time model to its continuous-time model. Illustrative example is given to demonstrate the effectiveness of the developed method.

Key Words : time-delay, Hankel matrix, Singular Value Decomposition, Generalized Bilinear Transformation

I. Introduction

Time delay is one of the key factors influencing the overall system stability and performance. In particular, as the different effects of actuator, sensor and controller exist in control systems, delays are often formulated as state time delays, input time delays as

well as output time delays in a continuous-time or discrete-time framework [1], [2-4]. To digitally simulate and design a continuous-time delayed control system, it is often required to obtain an equivalent discrete-time model. The digital modeling of continuous-time systems with input delays can be found in a standard textbook [5]. For improving the

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performance of a continuous-time system with multiple time delays, several advanced control theories and practical design techniques have been proposed [6-8]. Recently, a discretization method via the Chebyshev quadrature formula together with a linear interpolation method to construct an equivalent discrete-time model from the continuous time multiple time-delayed system was developed in [9]. Despite the significant progress that has been made on continuous/discrete time systems with multiple time delays, yet the digital modeling of a continuous-time system with multiple fractional/integer time delays in state, input and output is far from fully explored [4]. In this paper, we propose a new approximated discretization method and a generalized bilinear transformation method.

II. Modeling of Continuous-Time System using Generalized Bilinear Transformation

1. Problem formulation and model transformation

Consider a controllable, observable and stable [3] continuous-time multiple-input, multiple-output (MIMO) system with multiple state, input and output time delays described by

$$\begin{aligned}\dot{x}(t) &= \sum_{i=0}^J A_i x(t-\delta_i) + \sum_{i=0}^K B_i u(t-\gamma_i), \\ y(t) &= \sum_{i=0}^M C_i x(t-\zeta_i)\end{aligned}\quad (1)$$

where $x(t) \in R^n$ is the state with $x(t) = \phi(t)$ for $t \in [t_0 - \max\{\delta_i, \zeta_i\}, t_0]$, $u(t) \in R^m$ is the control input and $y(t) \in R^p$ is the output of the system. Here n is the order of the system, m is number of inputs and p is number of outputs of the system (1). Also, $\delta_i \geq 0$, $i = 0, 1, \dots, J$, are the state delays, $r_i \geq 0$, $i = 0, 1, \dots, K$, are the input delays and $\zeta_i \geq 0$, $i = 0, 1, \dots, M$, are the output delays. These delays can be fractional or integer multiple of the

sampling time T. The function $\phi(t) \in C[t_0 - \max\{\delta_i, \zeta_i\}, t_0]$ is a continuous vector-valued initial condition. The system matrices (A_i, B_i, C_i) are sets of real matrices defined with $A_i \in R^{n \times n}$, $i = 0, 1, \dots, J$, $B_i \in R^{n \times m}$, $i = 0, 1, \dots, K$, $C_i \in R^{p \times n}$, $i = 0, 1, \dots, M$.

The discrete-time model is shown in (2) as follows

$$\begin{aligned}x(kT+T) &= Gx(kT) + Hu(kT) \\ y(kT) &= \hat{C}x(kT)\end{aligned}\quad (2)$$

$$\begin{aligned}\text{where } G &= \sum_{\tilde{n}}^{-1/2} R_{\tilde{n}}^T H_1 S_{\tilde{n}} \sum_{\tilde{n}}^{-1/2}, \\ H &= \sum_{\tilde{n}}^{-1/2} S_{\tilde{n}}^T E_m, \quad E_m^T = [I_m \mathbf{0} \dots \mathbf{0}], \text{ and} \\ \hat{C} &= \beta C, \quad C = E_p^T R_{\tilde{n}} \sum_{\tilde{n}}^{1/2}, \quad E_p^T = [I_p \mathbf{0} \dots \mathbf{0}], \\ \sum_{\tilde{n}} &= \text{diag}\{\sigma_i, \dots, \sigma_{\tilde{n}}\}.\end{aligned}$$

β is a modification factor which can adjust the steady-state value of the system (2) to match the original continuous-time system (1).

After applying linear transformation, we get a Discrete-time Delay Difference Equation (DDDE) with integer delays as described in expressions shown below:

$$\begin{aligned}x_{01}(kT+T) &= \hat{G}_0 x_{01}(kT) + \hat{G}_{01} x_{01}(kT-T) \\ &\quad + \hat{H}_0 u(kT) + \hat{H}_{01} u(kT-T), \\ y(kT) &= \hat{C}_0 x_{01}(kT) + \hat{C}_{01} x_{01}(kT-T) \\ &\quad + \hat{D}_{01} u(kT-T),\end{aligned}\quad (3)$$

where,

$$\begin{aligned}\hat{G}_0 &= -G_{01}, \quad \hat{G}_{01} = -G_{02}, \quad \hat{H}_0 = H_{01}, \quad \hat{H}_{01} = H_{02}, \\ \hat{C}_0 &= C_{01}, \quad \hat{C}_{01} = -C_{02} G_{02}, \quad \hat{D}_{01} = C_{02} H_{02}.\end{aligned}$$

2. Generalized Bilinear Transformation

From the discrete-time multiple integer time-delayed system represented by (3) we can obtain a continuous-time multiple integer time-delayed model via a generalized bilinear transformation method.

Continuous-time systems with non-integer time-delays are not convenient for conventional multi-variable controller designing procedures and hence it is advantageous to obtain an equivalent continuous-time model with integer time-delays. Also it is easy to design controller and observer for continuous-time integer delay systems for digital control of sampled data systems. The procedure for computation of a continuous-time integer delay model from a given discrete-time integer time-delay state equation is discussed as below. For simplicity and better understanding, we consider a simple example of given a discrete-time system with a single state delay and appropriate system dimensions as represented in (4).

$$\begin{aligned} x(kT+T) &= G_0x(kT) + G_1x(kT-T) \quad (4) \\ &\quad + H_0u(kT) \end{aligned}$$

where, $u(kT)$ is a piecewise-constant input signal.

We want to find an equivalent continuous-time model with integer delay as shown in (5).

$$\dot{x}(t) = A_0x(t) + A_1x(t-T) + B_0u(t) \quad (5)$$

where,

$$u(t) = u(kT) \text{ for } kT \leq t < (kT+T).$$

Integrating equation (5) with limits $t=kT$ to $t=kT+T$ gives:

$$\begin{aligned} \int_{kT}^{kT+T} \dot{x}(t) dt &= \int_{kT}^{kT+T} A_0x(\lambda) d\lambda \\ &\quad + A_1 \int_{kT}^{kT+T} x(\lambda-T) d\lambda + B_0 \int_{kT}^{kT+T} u(\lambda) d\lambda \\ &= A_0 \int_{kT}^{kT+T} x(\lambda) d\lambda + A_1 \int_{kT-T}^{kT} x(\lambda) d\lambda \\ &\quad + B_0 \int_{kT}^{kT+T} 1 d\lambda \quad u(kt) \end{aligned} \quad (6)$$

The first and second integral terms in the right-hand side of equation (6) can be approximated by

the generalized bilinear transformation method or a generalized trapezoidal-rule method as follows:

$$\begin{aligned} \int_{kT}^{kT+T} x(\lambda) d\lambda &= T[\alpha x(kT+T) + (1-\alpha)x(kT)] \\ \text{and} \\ \int_{kT-T}^{kT} x(\lambda) d\lambda &= T[\alpha x(kT) + (1-\alpha)x(kT-T)] \end{aligned} \quad (7)$$

where, $0 \leq \alpha \leq 1$. Note that when $\alpha = 0.5$, the generalized bilinear transformation method reduces to the standard bilinear transformation method. Substituting (6) and (7) into (5) gives:

$$\begin{aligned} x(kT+T) - x(kT) &= \\ A_0[\alpha x(kT+T) + (1-\alpha)x(kT)]T \\ + A_1[\alpha x(kT) + (1-\alpha)x(kT-T)]T + B_0Tu(kT) \end{aligned}$$

Comparing equations to find the expressions for G_0 , G_1 , and H_0 as

$$\begin{aligned} G_0 &= (I - \alpha A_0 T)^{-1} [I + (1 - \alpha) A_0 T + \alpha A_1 T], \quad (8) \\ G_1 &= (I - \alpha A_0 T)^{-1} [1 - \alpha] A_1 T, \\ H_0 &= (I - \alpha A_0 T)^{-1} B_0 T \end{aligned}$$

From above expressions, we now calculate the system matrices of the continuous-time system with integer delays for different values of α .

Remark 1: For $0 < \alpha < 1$, the system matrices for continuous-time system with integer delays in (5) are as shown below.

$$\begin{aligned} A_0 &= \frac{1}{T} \left\{ \left[G_0 - \left(\frac{\alpha}{1-\alpha} \right) G_1 - I \right] [\alpha G_0 + (1-\alpha)I - \left(\frac{\alpha^2}{1-\alpha} \right) G_1]^{-1} \right\}, \\ A_1 &= \frac{1}{T} \left\{ \left(\frac{1}{1-\alpha} \right) [I - \alpha A_0 T] G_1 \right\}, \\ B_0 &= \frac{1}{T} [(I - \alpha A_0 T) H_0]. \end{aligned} \quad (9)$$

Remark 2: If $\alpha=0.5$, then the system matrices in (9) reduces to standard bilinear transform as shown in equations below.

$$\begin{aligned}
 A_0 &= -\left(\frac{2}{T}\right)[I_2 - (G_0 - G_1)][I_2 + (G_0 - G_1)]^{-1}, \\
 A_1 &= \frac{2}{T}[I_2 - (A_0(\frac{T}{2}))]G_1, \\
 B_0 &= \frac{1}{T}[I_2 - (A_0(\frac{T}{2}))]H_0.
 \end{aligned} \tag{10}$$

The expressions represented in (10) are expressions for the system matrices of the continuous-time system with integer delays.

III. Simulation and Results

A digital model obtained by the proposed SVD approach is compared with the digital model designed using the bilinear transformation method [9]. In this example, we discuss the accuracy of the proposed method (SVD approach) over the previous method of digital modeling (bilinear transformation). For comparison, we consider the following continuous time system with a state delay as described by

$$\begin{aligned}
 \dot{x}(t) &= A_0x(t) + A_1x(t - T_x) + Bu(t), \\
 y(t) &= Cx(t)
 \end{aligned} \tag{11}$$

where,

$$A_0 = \begin{bmatrix} 0 & 1 \\ -1 & -2 \end{bmatrix}, A_1 = \begin{bmatrix} 0 & 0 \\ 0.2 & 0.1 \end{bmatrix}, B = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, C = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix},$$

the delay time $T_x = 0.2s$, and the sampling period is $T = 0.15s$, $n = 2$.

Following the computation steps, let $E_{\sigma, \tilde{n}} < 1\%$,

the singular value matrix $\sum_{\tilde{n}}$ is obtained as follows.

$$\sum_{\tilde{n}} = diag\{0.7884 \quad 0.3177 \quad 0.0012 \quad 0.0009\}.$$

So the appropriate order for the digital model is $\tilde{n} = 2n$, and the extended discrete-time model (2) is obtained as

$$\begin{aligned}
 G &= \begin{bmatrix} 0.9597 & 0.0492 & 0.0057 & 0.0001 \\ -0.1285 & 0.7745 & 0.0341 & 0.0003 \\ 0.0076 & 0.0387 & 0.0831 & -0.0926 \\ -0.0001 & -0.0003 & 0.0927 & -0.0923 \end{bmatrix}, \\
 H &= \begin{bmatrix} 0.2508 \\ 0.3357 \\ -0.0254 \\ 0.0002 \end{bmatrix}, \\
 \hat{C} &= \begin{bmatrix} 0.2282 & -0.1405 & -0.0029 & -0.0001 \\ 0.0822 & 0.3213 & -0.0276 & -0.0002 \end{bmatrix}.
 \end{aligned}$$

The observable canonical form transformation matrix T_0 is picked as $T_0 = [G \bar{C}_0^T \ C_0^T]$,

$$\text{where } \bar{C}_0 = \begin{bmatrix} \hat{C}G \\ \hat{C} \end{bmatrix} \tilde{C}_0^T, \quad \tilde{C}_0 = [I_2 \ 0_2], \text{ so that the}$$

output vector $y(kT)$ is equal to the state vector $x_{01}(kT)$.

So the matrices of the integer time-delayed discrete-time model are

$$\begin{aligned}
 \hat{G}_0 &= \begin{bmatrix} -2.6361 & 0.4752 \\ -38.1408 & 4.3611 \end{bmatrix}, \\
 \hat{G}_{01} &= \begin{bmatrix} 3.6344 & 0.2155 \\ 38.1043 & 2.2601 \end{bmatrix}, \\
 \hat{H}_0 &= [0.0101 \quad -0.0080], \\
 \hat{H}_{01} &= [0.1292 \quad -0.00835], \\
 \hat{C}_0 &= \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}, \quad \hat{C}_{01} = 0_2, \quad \hat{D}_{01} = 0_2.
 \end{aligned}$$

The equivalent digital model of the given continuous-time system with a state delay represented by (11), is calculated based on the bilinear transformation method as

$$\begin{aligned}
 x_b(kT+T) &= G_{b0}x_b(kT) + G_{b1}x_b(kT-T) \\
 &\quad + G_{b2}x_b(kT-2T) + H_{b0}u(kT) \\
 y_b(kT) &= x_b(kT)
 \end{aligned} \tag{12}$$

with,

$$\begin{aligned}
 G_{b0} &= \begin{bmatrix} 0.9901 & 0.1293 \\ -0.1231 & 0.7346 \end{bmatrix}, \quad G_{b1} = \begin{bmatrix} 0.0015 & 0.0008 \\ 0.0185 & 0.0093 \end{bmatrix}, \\
 G_{b2} &= \begin{bmatrix} 0.0002 & 0.0001 \\ 0.0013 & 0.0006 \end{bmatrix}, \quad H_{b0} = \begin{bmatrix} 0.0102 \\ 0.1291 \end{bmatrix}.
 \end{aligned}$$

In Fig. 1, the state responses of the continuous-time system represented by (11), the digital model using the proposed method, described by (2), and the digital model determined by the bilinear transformation, represented in (12) are compared. The output response by the proposed method is more accurate than the bilinear transformation. Fig. 2 shows enlargement of a part of step responses of the continuous-time system and the digital model using the proposed method.

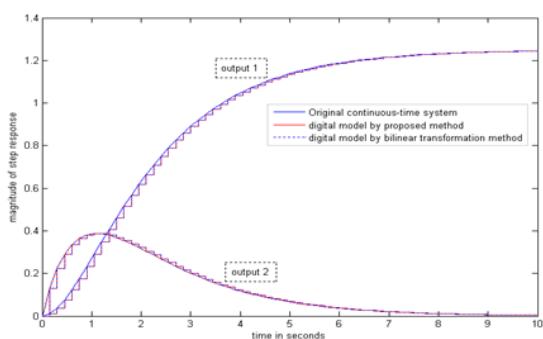


Fig. 1 Comparison of Step responses
그림 1. 계단 응답의 비교

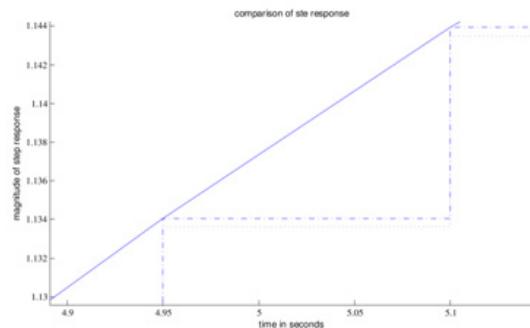


Fig. 2 Enlargement of a part of Step responses
그림 2. 계단 응답의 부분 확대

IV. Conclusion

In this paper, a new approximated state-space discretization scheme for a multivariable continuous-time system with multiple state, input and output delays has been presented. And a generalized bilinear

transform method for a multiple integer time-delayed systems is also proposed. As a result, the infinite-dimensional continuous-time control system can be converted into a finite dimensional sampled-data system, and a direct digital design of the sampled-data closed-loop system can be adopted.

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